

FEATURES

- **Maximum Offset Voltage: 70 μ V (25°C)**
- **Maximum Offset Drift: 0.8 μ V/°C**
- **Maximum Input Bias: 1pA (25°C) 40pA ($T_A \leq 85^\circ\text{C}$)**
- Open Loop Voltage Gain: 120dB Typ
- Gain Bandwidth Product: 3.6MHz
- CMRR: 100dB Min
- PSRR: 98dB Min
- 0.1Hz to 10Hz Noise: 1.3 μ V_{P-P}
- Supply Current: 330 μ A
- Rail-to-Rail Inputs and Outputs
- Unity Gain Stable
- 2.7V to 5.5V Operation Voltage
- Dual LTC6081 in 8-Lead MSOP and 10-Lead DFN10 Packages; Quad LTC6082 in 16-Lead SSOP and DFN Packages

APPLICATIONS

- Photodiode Amplifier
- Strain Gauge
- High Impedance Sensor Amplifier
- Microvolt Accuracy Threshold Detection
- Instrumentation Amplifiers
- Thermocouple Amplifiers

DESCRIPTION

The LTC®6081/LTC6082 are dual/quad low offset, low drift, low noise CMOS operational amplifiers with rail-to-rail input/output swing.

The 70 μ V maximum offset, 1pA input bias current, 120dB open loop gain and 1.3 μ V_{P-P} 0.1Hz to 10Hz noise make it perfect for precision signal conditioning. The LTC6081/LTC6082 features 100dB CMRR and 98dB PSRR.

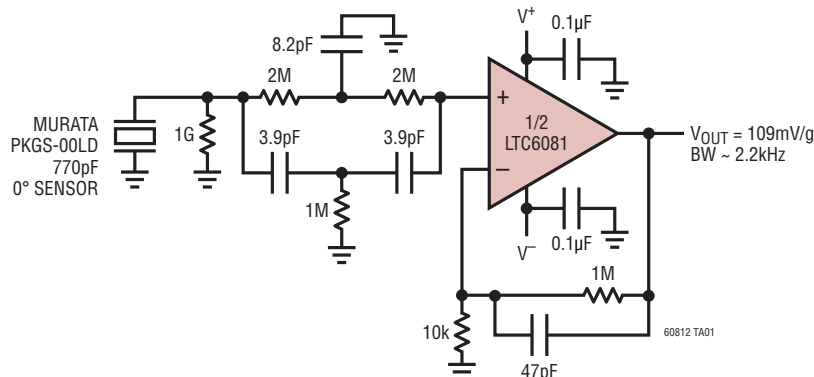
Each amplifier consumes only 330 μ A of current on a 3V supply. The 10-lead DFN has an independent shutdown function that reduces each amplifier's supply current to 1 μ A.

LTC6081/LTC6082 is specified for power supply voltages of 3V and 5V from -40°C to 125°C. The dual LTC6081 is available in 8-lead MSOP and 10-lead DFN10 packages. The quad LTC6082 is available in 16-lead SSOP and DFN packages.

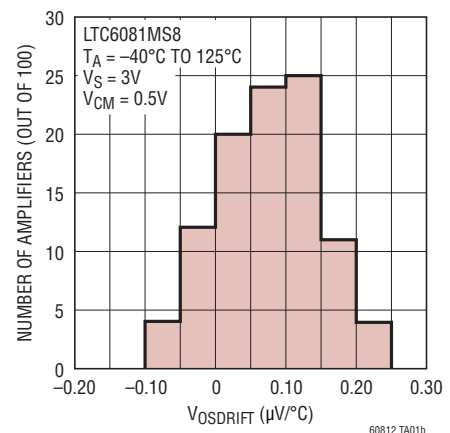
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TYPICAL APPLICATION

Shock Sensor Amplifier (Accelerometer)



V_{OS} Drift Histogram



60812fc

LTC6081/LTC6082

ABSOLUTE MAXIMUM RATINGS

(Note 1)

Total Supply Voltage (V^+ to V^-)	6V
Input Voltage	V^- to V^+
Output Short Circuit Duration (Note 2)	Indefinite
Operating Temperature Range (Note 3)	
LTC6081C, LTC6082C	-40°C to 85°C
LTC6081I, LTC6082I	-40°C to 85°C
LTC6081H, LTC6082H	-40°C to 125°C
(H Temperature Range Not Available in DFN Package)	

Specified Temperature Range (Note 4)	
LTC6081C, LTC6082C	0°C to 70°C
LTC6081I, LTC6082I	-40°C to 85°C
LTC6081H, LTC6082H	-40°C to 125°C
Junction Temperature	
DFN Packages	125°C
All Other Packages	150°C
Storage Temperature Range	
DFN Packages	-65°C to 125°C
All Other Packages	-65°C to 150°C
Lead Temperature (Soldering, 10 Sec)	300°C

PIN CONFIGURATION

TOP VIEW DD PACKAGE 10-LEAD (3mm × 3mm) PLASTIC DFN $T_{JMAX} = 125^{\circ}\text{C}$, $\theta_{JA} = 43^{\circ}\text{C/W}$ UNDERSIDE METAL CONNECTED TO V^-	TOP VIEW MS8 PACKAGE 8-LEAD PLASTIC MSOP $T_{JMAX} = 150^{\circ}\text{C}$, $\theta_{JA} = 200^{\circ}\text{C/W}$
TOP VIEW DHC PACKAGE 16-LEAD (5mm × 3mm) PLASTIC DFN $T_{JMAX} = 125^{\circ}\text{C}$, $\theta_{JA} = 43^{\circ}\text{C/W}$ UNDERSIDE METAL CONNECTED TO V^-	TOP VIEW GN PACKAGE 16-LEAD PLASTIC SSOP $T_{JMAX} = 150^{\circ}\text{C}$, $\theta_{JA} = 110^{\circ}\text{C/W}$

ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE
LTC6081CDD#PBF	LTC6081CDD#TRPBF	LCJP	10-Lead (3mm × 3mm) Plastic DFN	0°C to 70°C
LTC6081IDD#PBF	LTC6081IDD#TRPBF	LCJP	10-Lead (3mm × 3mm) Plastic DFN	–40°C to 85°C
LTC6081CMS8#PBF	LTC6081CMS8#TRPBF	LTCJN	8-Lead Plastic MSOP	0°C to 70°C
LTC6081IMS8#PBF	LTC6081IMS8#TRPBF	LTCJN	8-Lead Plastic MSOP	–40°C to 85°C
LTC6081HMS8#PBF	LTC6081HMS8#TRPBF	LTCJN	8-Lead Plastic MSOP	–40°C to 125°C
LTC6082CDHC#PBF	LTC6082CDHC#TRPBF	6082	16-Lead (5mm × 3mm) Plastic DFN	0°C to 70°C
LTC6082IDHC#PBF	LTC6082IDHC#TRPBF	6082	16-Lead (5mm × 3mm) Plastic DFN	–40°C to 85°C
LTC6082CGN#PBF	LTC6082CGN#TRPBF	6082	16-Lead Plastic SSOP	0°C to 70°C
LTC6082IGN#PBF	LTC6082IGN#TRPBF	6082I	16-Lead Plastic SSOP	–40°C to 85°C
LTC6082HGN#PBF	LTC6082HGN#TRPBF	6082H	16-Lead Plastic SSOP	–40°C to 125°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. *The temperature grade is identified by a label on the shipping container. Consult LTC Marketing for information on non-standard lead based finish parts.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreeel/>

LTC6081/LTC6082

ELECTRICAL CHARACTERISTICS The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. Test conditions are $V^+ = 3\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = 0.5\text{V}$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	C, I SUFFIXES			H SUFFIX			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Offset Voltage	LTC6081MS8, LTC6082GN $V_{\text{CM}} = 0.5\text{V}, 2.5\text{V}$	●	–70	70	–70	70		μV
		LTC6081MS8, LTC6082GN $V_{\text{CM}} = 0.5\text{V}, 2.5\text{V}$	●	–90	90	–90	90		μV
		LTC6081DD, LTC6082DHC $V_{\text{CM}} = 0.5\text{V}, 2.5\text{V}$	●	–70	70				μV
		LTC6081DD, LTC6082DHC $V_{\text{CM}} = 0.5\text{V}, 2.5\text{V}$	●	–90	90				μV
$\Delta V_{\text{OS}}/\Delta T$	Input Offset Voltage Drift (Note 5)		●	±0.2	±0.8		±0.2	±0.8	$\mu\text{V}/^\circ\text{C}$
I_{B}	Input Bias Current (Note 6)		●	0.2	1 40		0.2	1 500	pA pA
I_{OS}	Input Offset Current		●	0.1	15		0.1	100	pA pA
e_{n}	Input Referred Noise	Noise Density at $f = 1\text{kHz}$ Integrated Noise From 0.1Hz to 10Hz		13 1.3			13 1.3		nV/ $\sqrt{\text{Hz}}$ $\mu\text{V}_{\text{P-P}}$
I_{n}	Input Noise Current Density (Note 7)			0.5			0.5		fA/ $\sqrt{\text{Hz}}$
	Input Common Mode Range		●	V^-	V^+	V^-	V^+		V
C_{DIFF}	Differential Input Capacitance			3			3		pF
C_{CM}	Common Mode Input Capacitance			7			7		pF
CMRR	Common Mode Rejection Ratio	$V_{\text{CM}} = 0\text{V}$ to 1.5V	●	95	105		95	105	dB
		$V_{\text{CM}} = 0\text{V}$ to 1.5V	●	88	100		86	100	dB
		$V_{\text{CM}} = 0\text{V}$ to 3V	●	93	105		93	105	dB
		$V_{\text{CM}} = 0\text{V}$ to 3V	●	88	100		86	100	dB
PSRR	Power Supply Rejection Ratio	$V_{\text{S}} = 2.7\text{V}$ to 5.5V	●	98	110		98	110	dB
			●	96			96		dB
V_{OUT}	Output Voltage, High, Either Output Pin	No Load	●		1			1	mV
		$I_{\text{SOURCE}} = 0.5\text{mA}$	●	–32			–35		mV
		$I_{\text{SOURCE}} = 5\text{mA}$	●	–320			–350		mV
	Output Voltage, Low, Either Output Pin (Referred to V^-)	No Load	●		1			1	mV
		$I_{\text{SINK}} = 0.5\text{mA}$	●		33			40	mV
		$I_{\text{SINK}} = 5\text{mA}$	●		300			360	mV
A_{VOL}	Large-Signal Voltage Gain	$R_{\text{LOAD}} = 10\text{k}$, $0.5\text{V} < V_{\text{OUT}} < 2.5\text{V}$	●	110	120		110	120	dB
I_{SC}	Output Short-Circuit Current	Source	●	17			15		mA
		Sink	●	17			15		mA
SR	Slew Rate	$A_{\text{V}} = 1$			1			1	V/ μs
GBW	Gain-Bandwidth Product ($f_{\text{TEST}} = 50\text{kHz}$)	$R_{\text{L}} = 100\text{k}$	●	2.5	3.6		2.5	3.6	MHz
				1.8			1.5		MHz
Φ_0	Phase Margin	$R_{\text{L}} = 10\text{k}$			70			70	Deg
t_{S}	Settling Time 0.1%	$A_{\text{V}} = 1$, 1V Step			6			6	μs
I_{S}	Supply Current (Per Amplifier)	No Load	●		330 400 435			330 400 460	μA μA
			●		0.5 2				μA μA
V_{S}	Supply Voltage Range	Guaranteed by the PSRR Test	●	2.7	5.5		2.7	5.5	V
	Channel Separation	$f_{\text{S}} = 10\text{kHz}$, $R_{\text{L}} = 10\text{k}$			–120			–120	dB

ELECTRICAL CHARACTERISTICS The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. Test conditions are $V^+ = 3\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = 0.5\text{V}$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		C, I SUFFIXES			H SUFFIX			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	
	Shutdown Logic	$\overline{\text{SHDN}}$ High $\overline{\text{SHDN}}$ Low	● ●	2		0.8	2		0.8	V V
THD	Total Harmonic Distortion	$f = 10\text{kHz}$, $V^+ = 3\text{V}$, $V_{\text{OUT}} = 1\text{V}_{\text{P-P}}$, $R_L = 10\text{k}$			-90			-90		dB
t_{ON}	Turn-On Time	$V_{\overline{\text{SHDN}}} = 0.8\text{V}$ to 2V			10			10		μs
t_{OFF}	Turn-Off Time	$V_{\overline{\text{SHDN}}} = 2\text{V}$ to 0.8V			2			2		μs
	$\overline{\text{SHDN}}$ Pin Current	$V_{\overline{\text{SHDN}}} = 0\text{V}$	●			2				μA

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. Test conditions are $V^+ = 5\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = 0.5\text{V}$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		C, I SUFFIXES			H SUFFIX			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Offset Voltage	LTC6081MS8, LTC6082GN $V_{\text{CM}} = 0.5\text{V}$	●	-70		70	-70		70	μV
		LTC6081MS8, LTC6082GN $V_{\text{CM}} = 0.5\text{V}$	●	-90		90	-90		90	μV
		LTC6081DD, LTC6082DHC $V_{\text{CM}} = 0.5\text{V}$	●	-70		70				μV
		LTC6081DD, LTC6082DHC $V_{\text{CM}} = 0.5\text{V}$	●	-90		90				μV
$\Delta V_{\text{OS}}/\Delta T$	Input Offset Voltage Drift (Note 8)		●		± 0.2	± 0.8		± 0.2	± 0.8	$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current		●		0.2	40		0.2	500	pA pA
I_{OS}	Input Offset Current		●		0.1	15		0.1	100	pA pA
e_n	Input Referred Noise	$f = 1\text{kHz}$ 0.1Hz to 10Hz			13 1.3			13 1.3		$\text{nV}/\sqrt{\text{Hz}}$ $\mu\text{V}_{\text{P-P}}$
I_n	Input Noise Current Density (Note 7)				0.5			0.5		$\text{fA}/\sqrt{\text{Hz}}$
	Input Common Mode Range		●	V^-		V^+	V^-		V^+	V
C_{DIFF}	Differential Input Capacitance				3			3		pF
C_{CM}	Common Mode Input Capacitance				7			7		pF
CMRR	Common Mode Rejection Ratio	$V_{\text{CM}} = 0\text{V}$ to 3.5V	●	100	110		100	110		dB
		$V_{\text{CM}} = 0\text{V}$ to 3.5V	●	95	110		94	110		dB
		$V_{\text{CM}} = 0\text{V}$ to 5V	●	86	95		86	95		dB
PSRR	Power Supply Rejection Ratio	$V_S = 2.7\text{V}$ to 5.5V	●	98	110		98	110		dB
			●	96			96			dB
V_{OUT}	Output Voltage, High, Either Output Pin (Referred to V^+)	No Load			1			1		mV
		$I_{\text{SOURCE}} = 0.5\text{mA}$	●	-24			-25			mV
		$I_{\text{SOURCE}} = 5\text{mA}$	●	-200			-220			mV
	Output Voltage, Low, Either Output Pin (Referred to V^-)	No Load			1			1		mV
		$I_{\text{SINK}} = 0.5\text{mA}$	●			27			32	mV
		$I_{\text{SINK}} = 5\text{mA}$	●			210			240	mV
A_{VOL}	Large-Signal Voltage Gain	$R_{\text{LOAD}} = 10\text{k}$, $0.5\text{V} < V_{\text{OUT}} < 4.5\text{V}$	●	110	120		110	120		dB

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. Test conditions are $V^+ = 5\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = 0.5\text{V}$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	C, I SUFFIXES			H SUFFIX			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
I_{SC}	Output Short-Circuit Current	Source Sink	● 24 ● 24			21 21			mA mA
SR	Slew Rate	$A_V = 1$		1			1		V/ μs
GBW	Gain-Bandwidth Product ($f_{\text{TEST}} = 50\text{kHz}$)	$R_L = 100\text{k}$	● 2.5 1.8	3.5		2.5 1.5	3.5		MHz MHz
Φ_0	Phase Margin	$R_L = 10\text{k}$		70			70		Deg
t_S	Settling Time 0.1%	$A_V = 1$, 1V Step		6			6		μs
I_S	Supply Current (Per Amplifier)	No Load	●	340	425 465		340	425 490	μA μA
	Shutdown Current (Per Amplifier)	Shutdown, $V_{\text{SHDN}} \leq 1.2\text{V}$	●		6				μA
V_S	Supply Voltage Range	Guaranteed by the PSRR Test	●	2.7	5.5	2.7		5.5	V
	Channel Separation	$f_s = 10\text{kHz}$, $R_L = 10\text{k}$		-120			-120		dB
	Shutdown Logic	SHDN High SHDN Low	● ●	3.5	1.2	3.5		1.2	V V
THD	Total Harmonic Distortion	$f = 10\text{kHz}$, $V^+ = 5\text{V}$, $V_{\text{OUT}} = 2V_{\text{P-P}}$, $R_L = 10\text{k}$		-90			-90		dB
t_{ON}	Turn-On Time	$V_{\text{SHDN}} = 1.2\text{V}$ to 3.5V		10			10		μs
t_{OFF}	Turn-Off Time	$V_{\text{SHDN}} = 3.5\text{V}$ to 1.2V		2			2		μs
	SHDN Pin Current	$V_{\text{SHDN}} = 0\text{V}$	●		2				μA

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: A heat sink may be required to keep the junction temperature below the absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted.

Note 3: The LTC6081C/LTC6082C and LTC6081I/LTC6082I are guaranteed functional over the operating temperature range of -40°C to 85°C . The LTC6081H/LTC6082H are guaranteed functional over the operating temperature range of -40°C to 125°C .

Note 4: The LTC6081C/LTC6082C are guaranteed to meet specified performance from 0°C to 70°C . The LTC6081C/LTC6082C are designed,

characterized and expected to meet specified performance from -40°C to 85°C but are not tested or QA sampled at these temperatures. The LTC6081I/LTC6082I are guaranteed to meet specified performance from -40°C to 85°C . The LTC6081H/LTC6082H are guaranteed to meet specified performance from -40°C to 125°C .

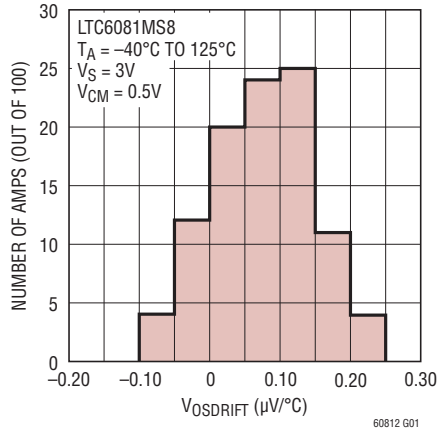
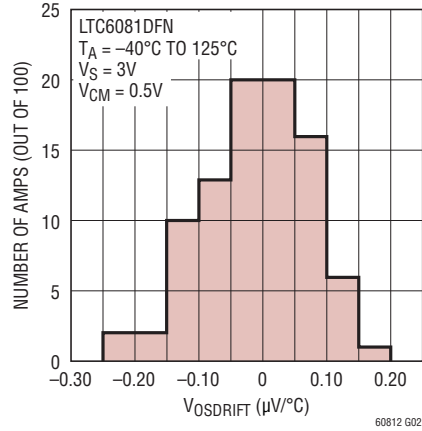
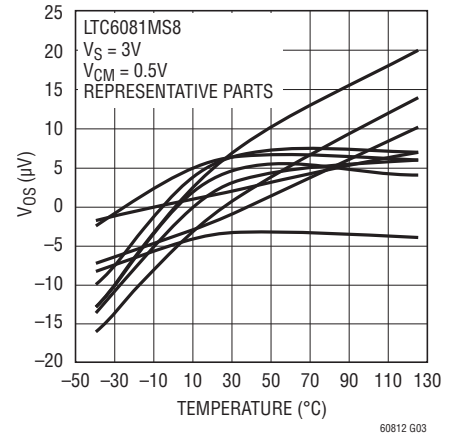
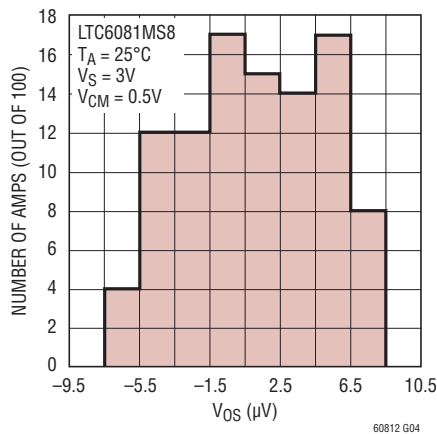
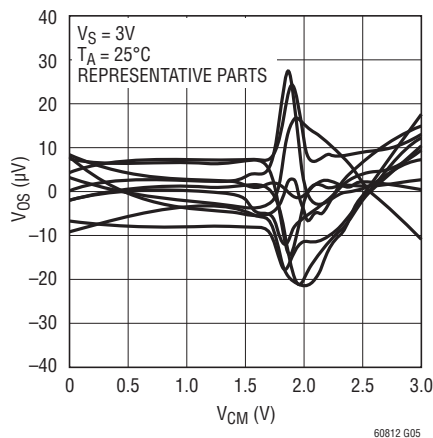
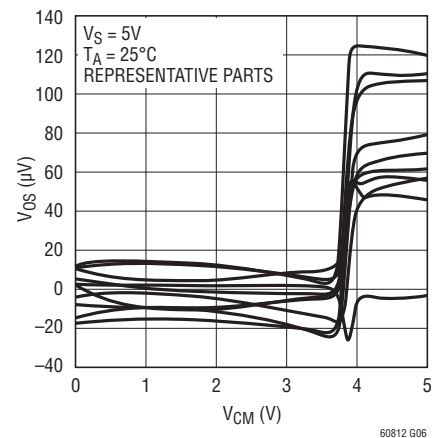
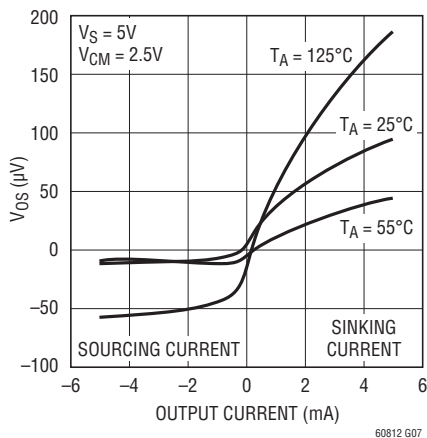
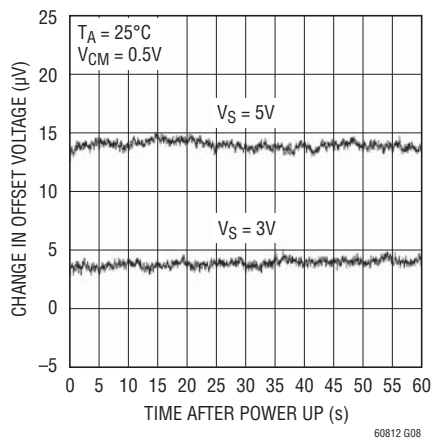
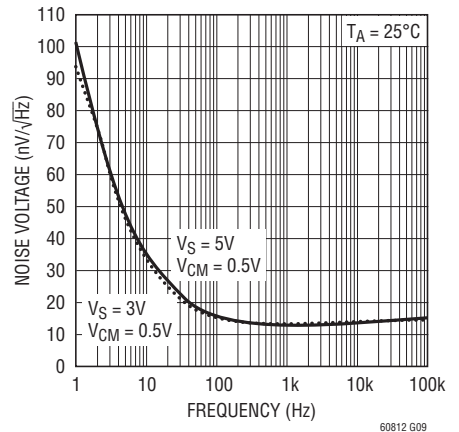
Note 5: Input offset drift is computed from the limits of the V_{OS} test divided by the temperature range. This is a conservative estimate of worst case drift. Consult the Typical Performance Characteristics section for more information on input offset drift.

Note 6: I_B guaranteed by the $V_S = 5\text{V}$ test.

Note 7: Current noise is calculated from $I_n = \sqrt{2qI_B}$, where $q = 1.6 \cdot 10^{-19}$ coulomb.

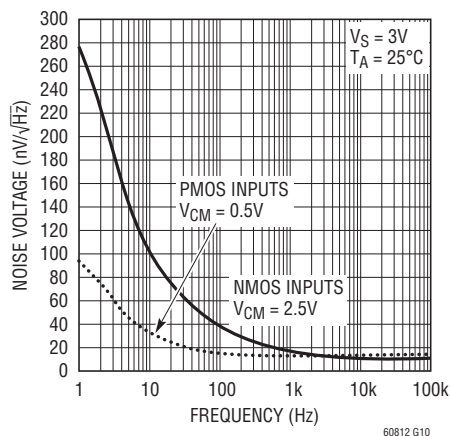
Note 8: V_{OS} drift is guaranteed by the $V_S = 3\text{V}$ test.

TYPICAL PERFORMANCE CHARACTERISTICS

V_{OS} Drift Histogram**V_{OS} Drift Histogram****V_{OS} vs Temperature****V_{OS} Histogram****V_{OS} vs V_{CM}****V_{OS} vs V_{CM}****V_{OS} vs Output Current****Warm-Up Drift vs Time****Noise Voltage vs Frequency**

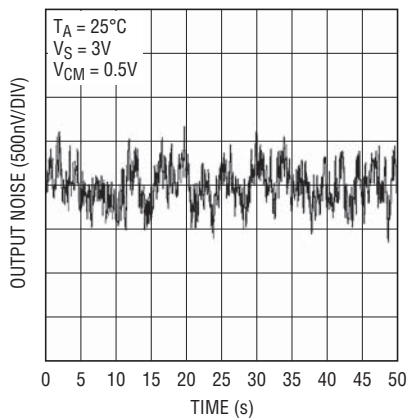
TYPICAL PERFORMANCE CHARACTERISTICS

Noise Voltage vs Frequency



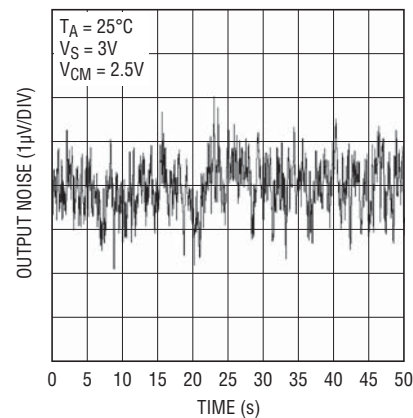
60812 G10

0.1Hz to 10Hz Output Voltage Noise



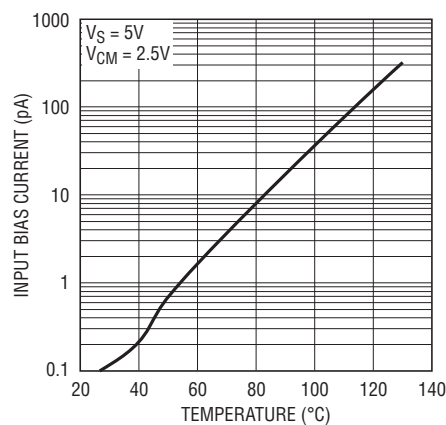
60812 G11

0.1Hz to 10Hz Output Voltage Noise

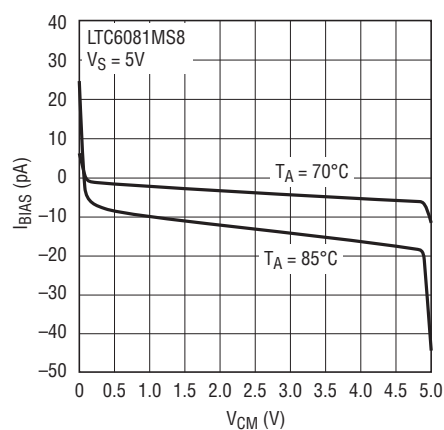


60812 G12

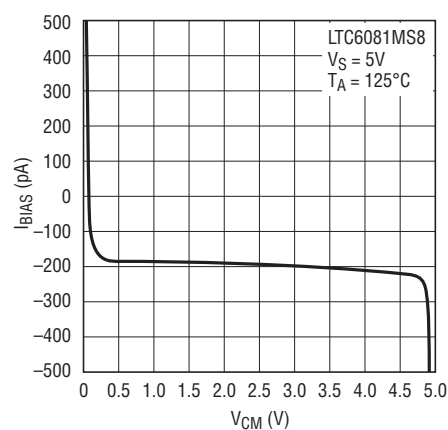
Input Bias Current vs Temperature



60812 G13

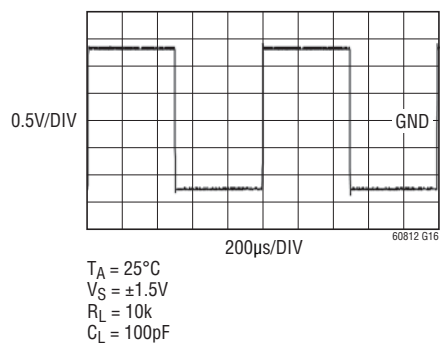
 I_{BIAS} vs V_{CM} 

60812 G14

 I_{BIAS} vs V_{CM} 

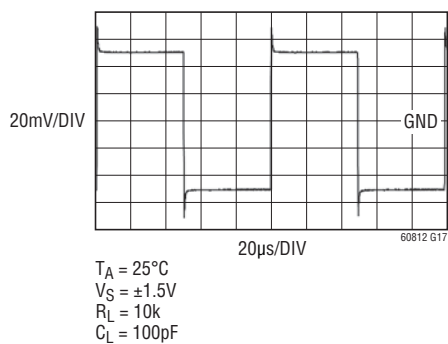
60812 G15

Large Signal Transient

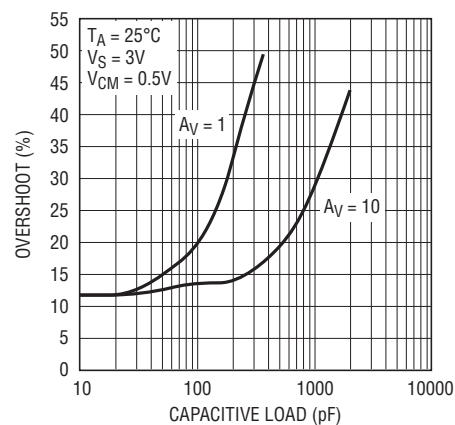


60812 G16

Small Signal Transient



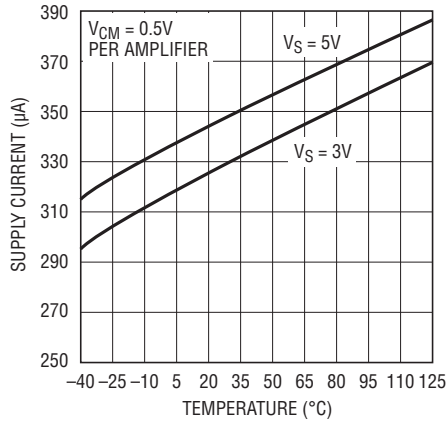
60812 G17

Overshoot vs C_L 

60812 G18

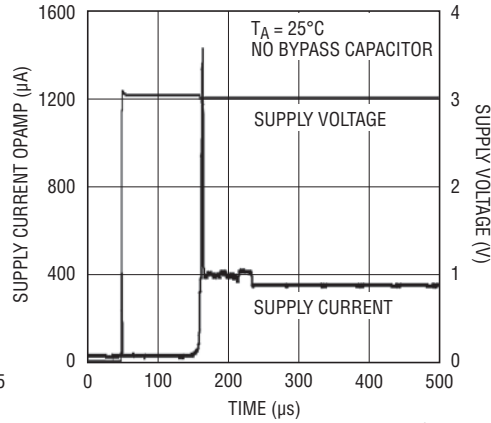
TYPICAL PERFORMANCE CHARACTERISTICS

Supply Current vs Temperature



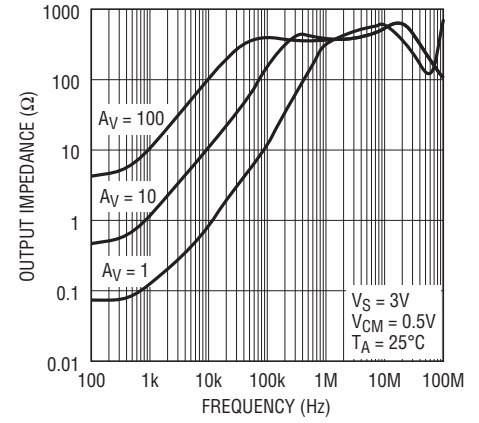
60812 G19

Supply Current vs Time



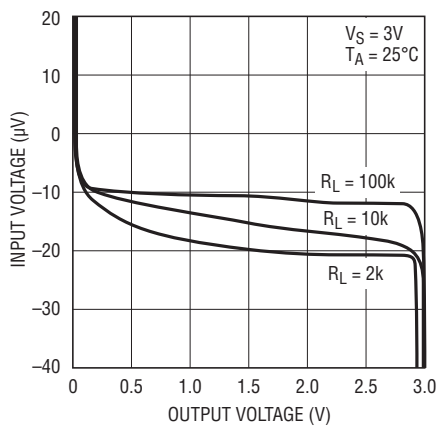
60812 G20

Output Impedance vs Frequency



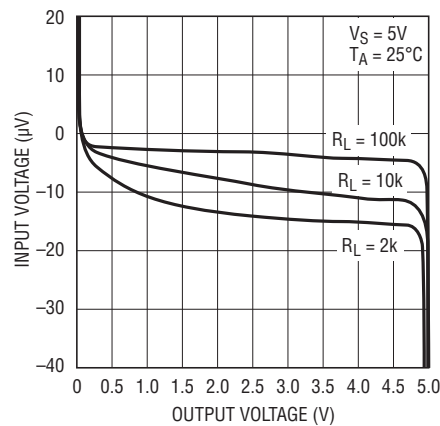
60812 G21

Open Loop Gain



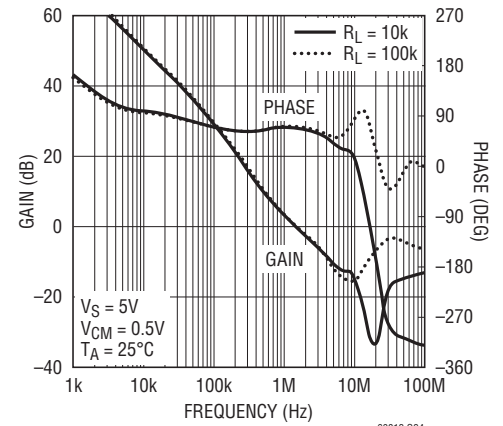
60812 G22

Open Loop Gain



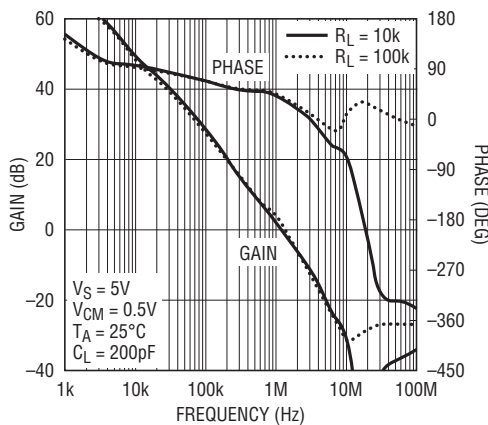
60812 G23

Open Loop Gain vs Frequency



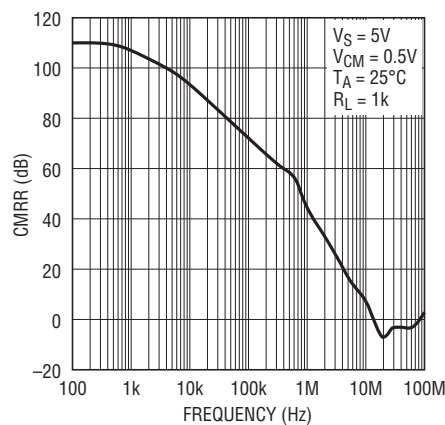
60812 G24

Open Loop Gain vs Frequency



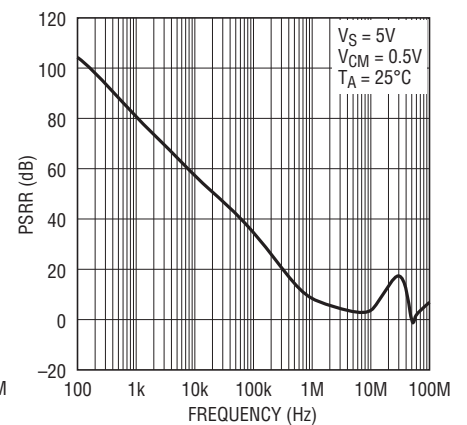
60812 G25

CMRR vs Frequency



60812 G26

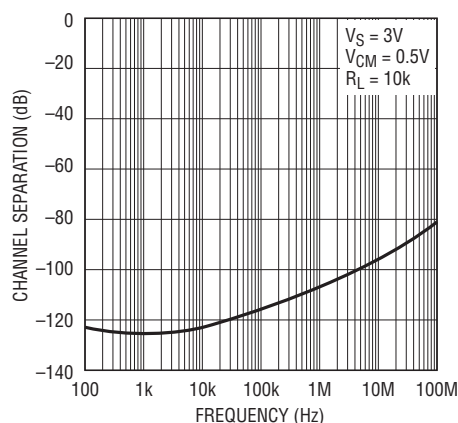
PSRR vs Frequency



60812 G27

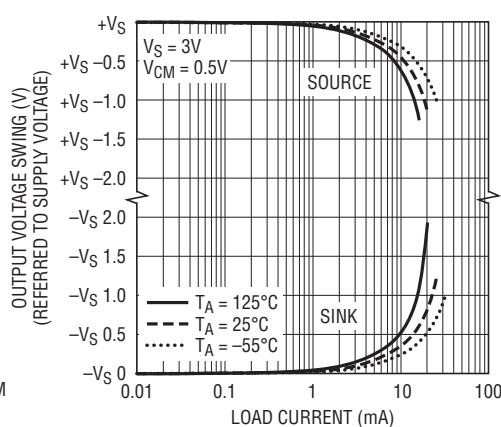
TYPICAL PERFORMANCE CHARACTERISTICS

Channel Separation vs Frequency



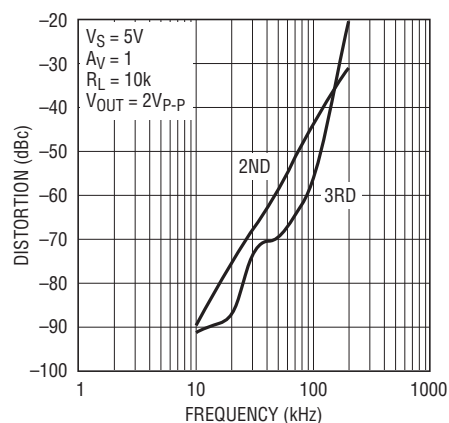
60812 G28

Output Voltage Swing vs Load Current



60789 G29

Distortion vs Frequency



60812 G30

PIN FUNCTIONS

OUT: Amplifier Output**-IN:** Inverting Input**+IN:** Noninverting Input**V⁺:** Positive Supply**V⁻:** Negative Supply

SHDN_A: Shutdown Pin of Amplifier A, active low and only valid for LTC6081DD. An internal current source pulls the pin to V^+ when floating.

SHDN_B: Shutdown Pin of Amplifier B, active low and only valid for LTC6081DD. An internal current source pulls the pin to V^+ when floating.

NC: Not internally connected.

Exposed Pad: Connected to V^- .

APPLICATIONS INFORMATION

Preserving Input Precision

Preserving input accuracy of the LTC6081/LTC6082 requires that the application circuit and PC board layout do not introduce errors comparable or greater than the $5\mu\text{V}$ typical offset of the amplifiers. Temperature differentials across the input connections can generate thermocouple voltages of 10's of microvolts so the connections to the input leads should be short, close together and away from heat dissipating components. Air current across the board can also generate temperature differentials.

The extremely low input bias currents (0.1pA typical) allow high accuracy to be maintained with high impedance sources and feedback resistors. Leakage currents on the PC board can be higher than the input bias current. For example, $10\text{G}\Omega$ of leakage between a 5V supply lead and an input lead will generate 500pA ! Surround the input leads with a guard ring driven to the same potential as the input common mode voltage to avoid excessive leakage in high impedance applications.

Capacitive Load

LTC6081/LTC6082 can drive capacitive load up to 200pF in unity gain. The capacitive load driving capability increases as the amplifier is used in higher gain configurations. A small series resistance between the output and the load further increases the amount of capacitance the amplifier can drive.

SHDN Pins

Pins 5 and 6 are used for power shutdown on the LTC6081 in the DD package. If they are floating, internal current sources pull Pins 5 and 6 to V^+ and the amplifiers operate normally. In shutdown, the amplifier output is high impedance, and each amplifier draws less than $2\mu\text{A}$ current.

Rail-to-Rail Input

The input stage of LTC6081/LTC6082 combines both PMOS and NMOS differential pairs, extending its input common mode voltage range to both positive and negative supply voltages. At high input common mode range, the NMOS pair is on. At low common mode range, the PMOS pair is on. The transition happens when the common voltage is between 1.3V and 0.9V below the positive supply. LTC6081 has better low frequency noise performance with PMOS input on due to its lower flicker noise (see Voltage Noise vs Frequency and 0.1Hz to 10Hz Input Voltage Noise in Typical Performance Characteristics).

Thermal Hysteresis

Figure 1 shows the input offset voltage hysteresis of the LTC6081MS8 for 3 thermal cycles from -45°C to 90°C . The typical offset shift is $\pm 4\mu\text{V}$. The data was taken with the ICs in stress free sockets. Mounting to PC boards may cause additional hysteresis due to mechanical stress. The LTC6081 will meet offset voltage specifications in the electrical characteristics table even after $15\mu\text{V}$ of additional error from thermal hysteresis.

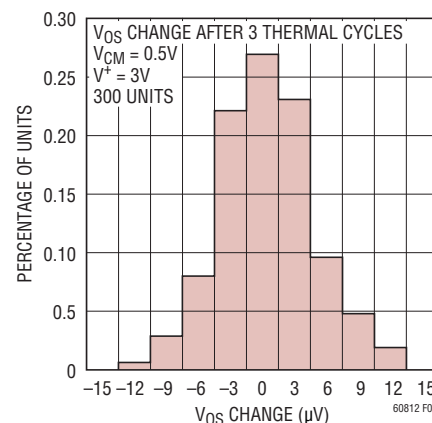


Figure 1. V_{OS} Thermal Hysteresis of LTC6081MS8

APPLICATIONS INFORMATION

PC Board Layout

Mechanical stress on a PC board and soldering-induced stress can cause the V_{OS} and V_{OS} drift to shift. The DD and DHC packages are more sensitive to stress. A simple way to reduce the stress-related shifts is to mount the IC near the short edge of the PC board, or in a corner. The board edge acts as a stress boundary, or a region where the flexure of the board is minimum. The package should always be mounted so that the leads absorb the stress

and not the package. The package is generally aligned with the leads perpendicular to the long side of the PC board (see Figure 2).

The most effective technique to relieve the PC board stress is to cut slots in the board around the op amp. These slots can be cut on three sides of the IC and the leads can exit on the fourth side. Figure 2 shows the layout of a LTC6081DD with slots at three sides.

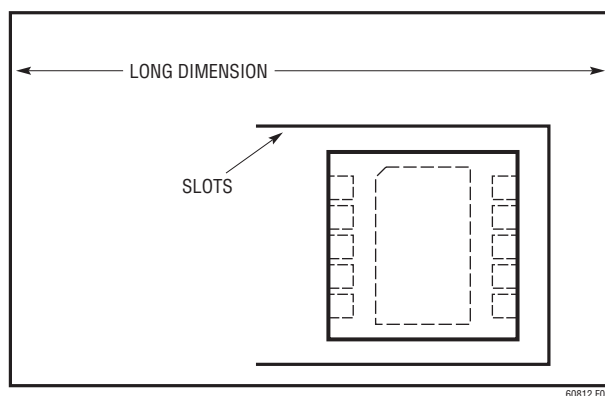
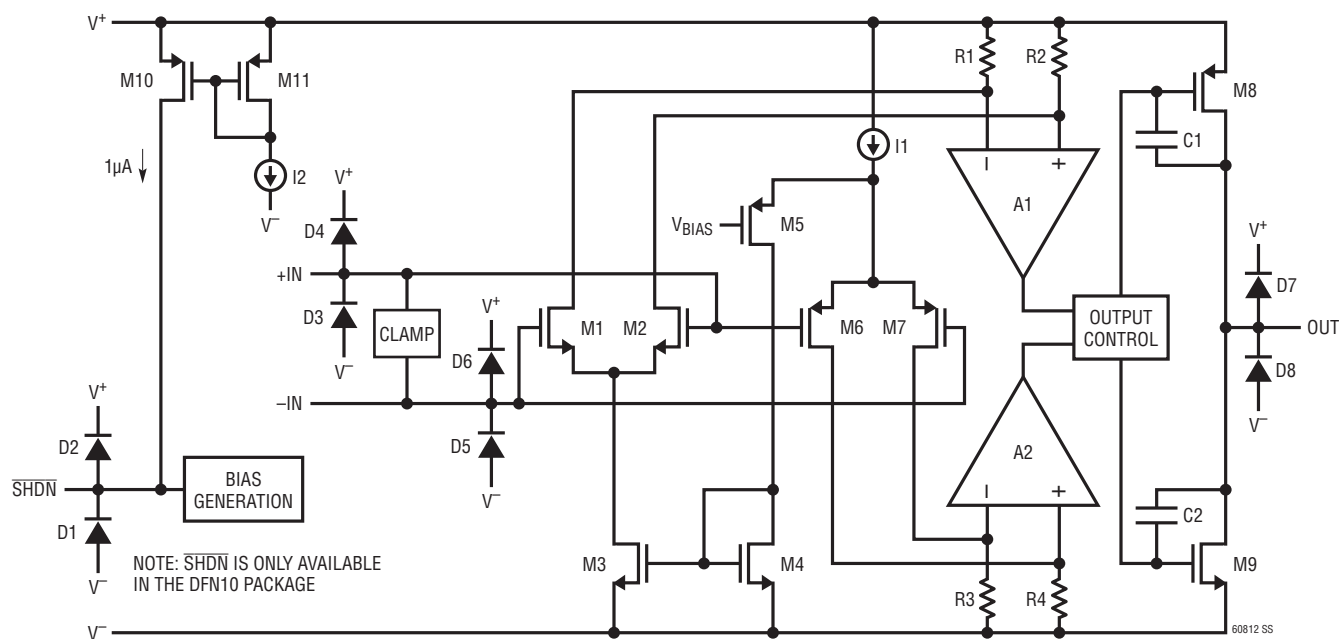


Figure 2. Vertical Orientation of LTC6081DD with Slots

SIMPLIFIED SCHEMATIC

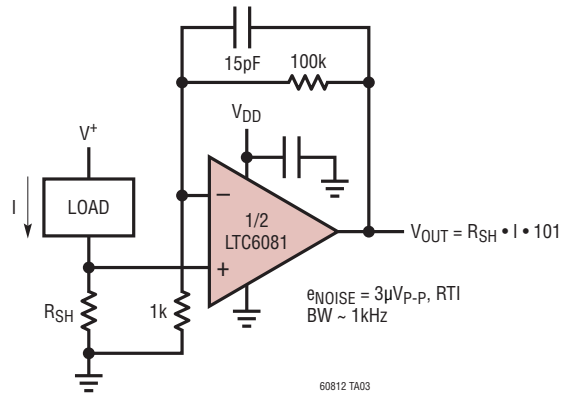
Simplified Schematic of the Amplifier



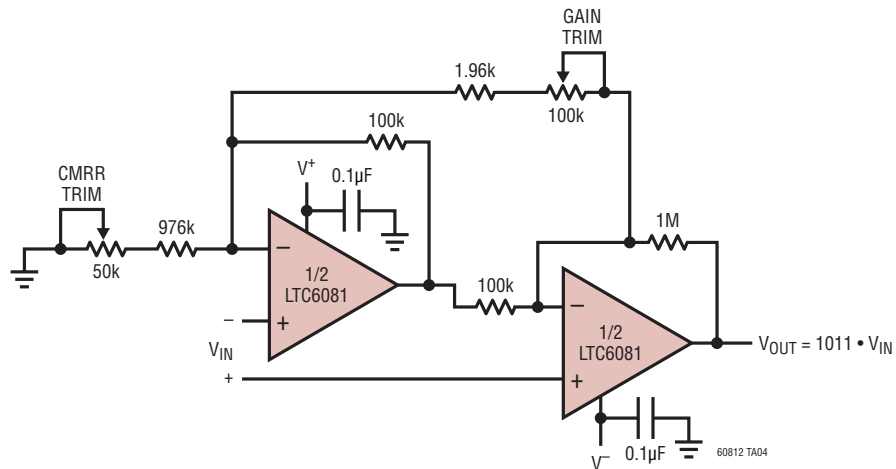
60812fc

TYPICAL APPLICATIONS

Low Side Current Sense



Two Op-Amp Instrumentation Amplifier



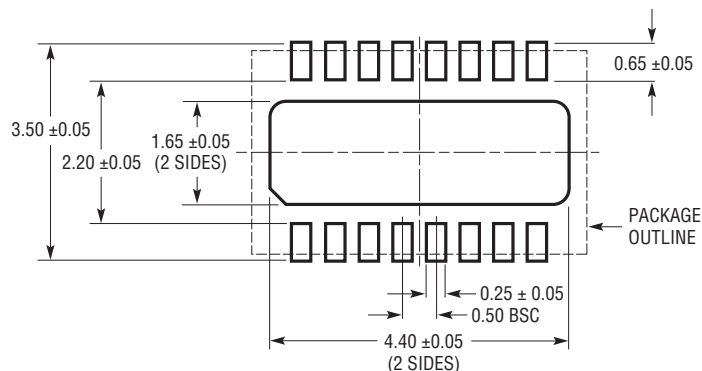
The circuit diagram shows an LTC6081 op-amp configured as a precision centiVolts-to-Volts converter. The non-inverting input (+) is connected to a 5V supply through a 1MΩ resistor and a 0.1μF capacitor. The inverting input (-) is connected to a 5V supply through a 1MΩ resistor and a 1μF capacitor. The output of the op-amp is connected to a 2.49MΩ resistor, which is then connected to a 100pF capacitor and a 10kΩ resistor to ground. The output voltage is labeled as $V_{OUT} = 10\text{mV}/^{\circ}\text{C}$ from 0°C to 500°C . The op-amp is labeled "1/2 LTC6081".

60812 TA05

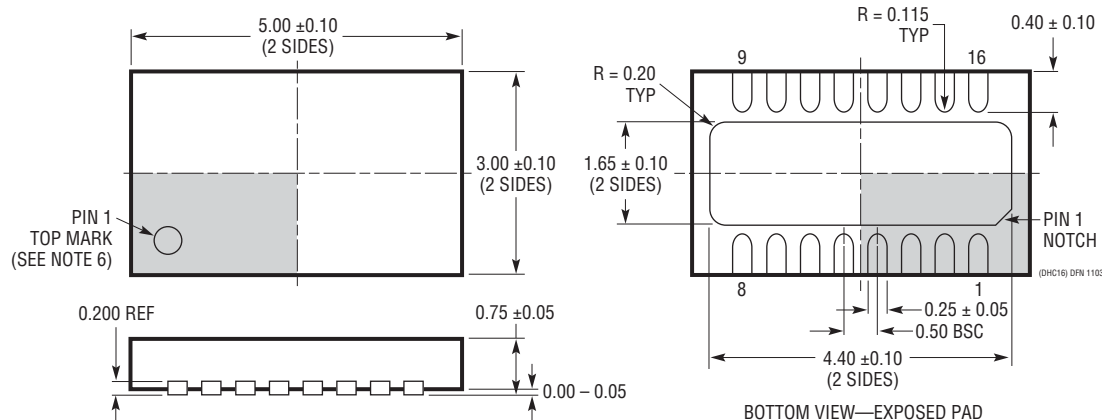
$I_{OUT} = -1\text{nA} \rightarrow 1\text{nA}$ FOR
 $V_{IN} = -10\text{V} \rightarrow 10\text{V}$
 TOTAL ERROR $< \pm 1\%$ (10pA)

PACKAGE DESCRIPTION

DHC Package
16-Lead Plastic DFN (5mm × 3mm)
 (Reference LTC DWG # 05-08-1706)



RECOMMENDED SOLDER PAD PITCH AND DIMENSIONS

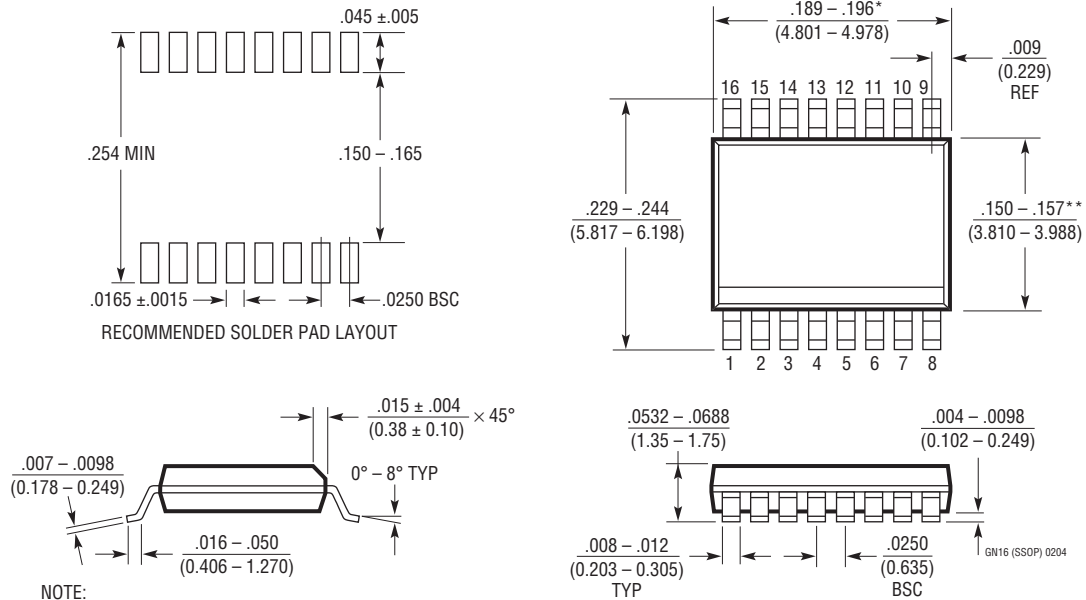


NOTE:

1. DRAWING PROPOSED TO BE MADE VARIATION OF VERSION (WJED-1) IN JEDEC PACKAGE OUTLINE MO-229
2. DRAWING NOT TO SCALE
3. ALL DIMENSIONS ARE IN MILLIMETERS
4. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE
5. EXPOSED PAD SHALL BE SOLDER PLATED
6. SHADED AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION ON THE TOP AND BOTTOM OF PACKAGE

PACKAGE DESCRIPTION

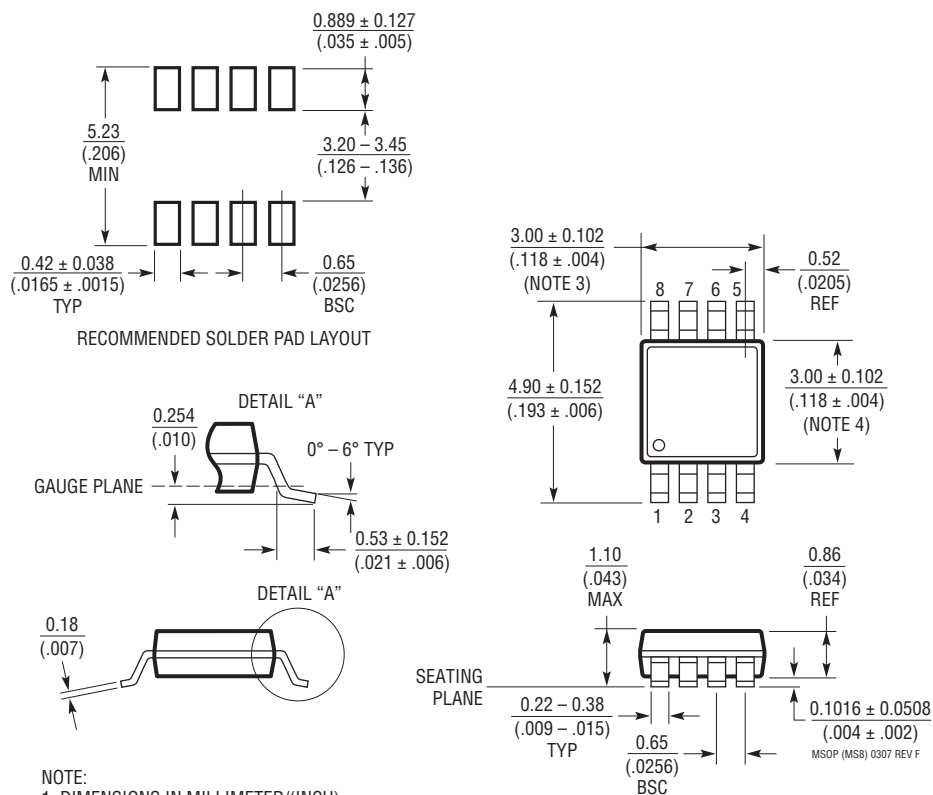
GN Package 16-Lead Plastic SSOP (Narrow .150 Inch) (Reference LTC DWG # 05-08-1641)



PACKAGE DESCRIPTION

MS8 Package
8-Lead Plastic MSOP

(Reference LTC DWG # 05-08-1660 Rev F)



NOTE:

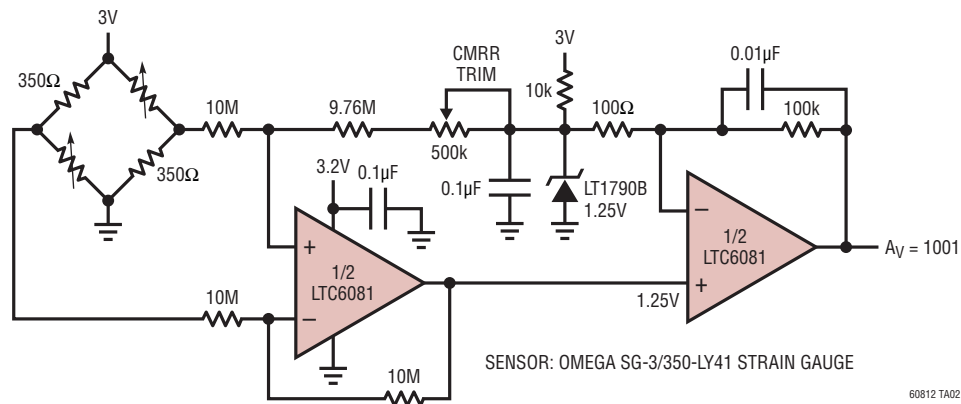
1. DIMENSIONS IN MILLIMETER/(INCH)
2. DRAWING NOT TO SCALE
3. DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.152mm (.006") PER SIDE
4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (.006") PER SIDE
5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (.004") MAX

REVISION HISTORY (Revision history begins at Rev B)

REV	DATE	DESCRIPTION	PAGE NUMBER
B	3/10	Change LT to LTC on all part numbers in Order Information Section.	3
C	07/10	Update to Simplified Schematic	12

TYPICAL APPLICATION

Single Supply Strain Gauge Amplifier



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1678/LT1679	Dual/Quad Precision Op Amps	Low Noise, 2.7V to 36V Operation
LTC2050	Zero-Drift Op Amp	2.7V Operation, SOT-23 Package
LTC2051/LTC2052	Dual/Quad Zero-Drift Op Amps	MS8/GN16 Packages
LTC2054/LTC2055	Single/Dual Zero-Drift Op Amp	Micropower, SOT-23 and DFN Packages
LTC6078/LTC6079	Dual/Quad Low Noise Precision CMOS Op Amps	Micropower 0.7μV/°C V _{OS} Drift
LTC6241/LTC6242	Dual/Quad Low Noise CMOS Op Amps	18MHz Bandwidth, 10V/μs Slew Rate
LTC6244	Dual 50MHz CMOS Op Amp	Low Noise, Rail-to-Rail Out, MS8 and DFN Packages